

CONSIDER THE COMMUNICATION PROBLEM \Rightarrow MINIMIZE P_e AND ASSUME $P(H_0) = P(H_1)$. WE HAVE AN ML RECEIVER OR DECIDE H_1 IF

$$\frac{p(x|H_1)}{p(x|H_0)} \geq \gamma = \frac{P(H_0)}{P(H_1)} = 1$$

BUT $p(x|H_i) = \frac{1}{(2\pi\sigma^2)^{N/2}} e^{-\frac{1}{2\sigma^2} \sum_{n=0}^{N-1} (x(n) - s_i(n))^2}$

CHOOSE H_i FOR WHICH

$$D_i^2 = \sum_{n=0}^{N-1} (x(n) - s_i(n))^2$$

IS MINIMUM. CALLED A MINIMUM DISTANCE RECEIVER.

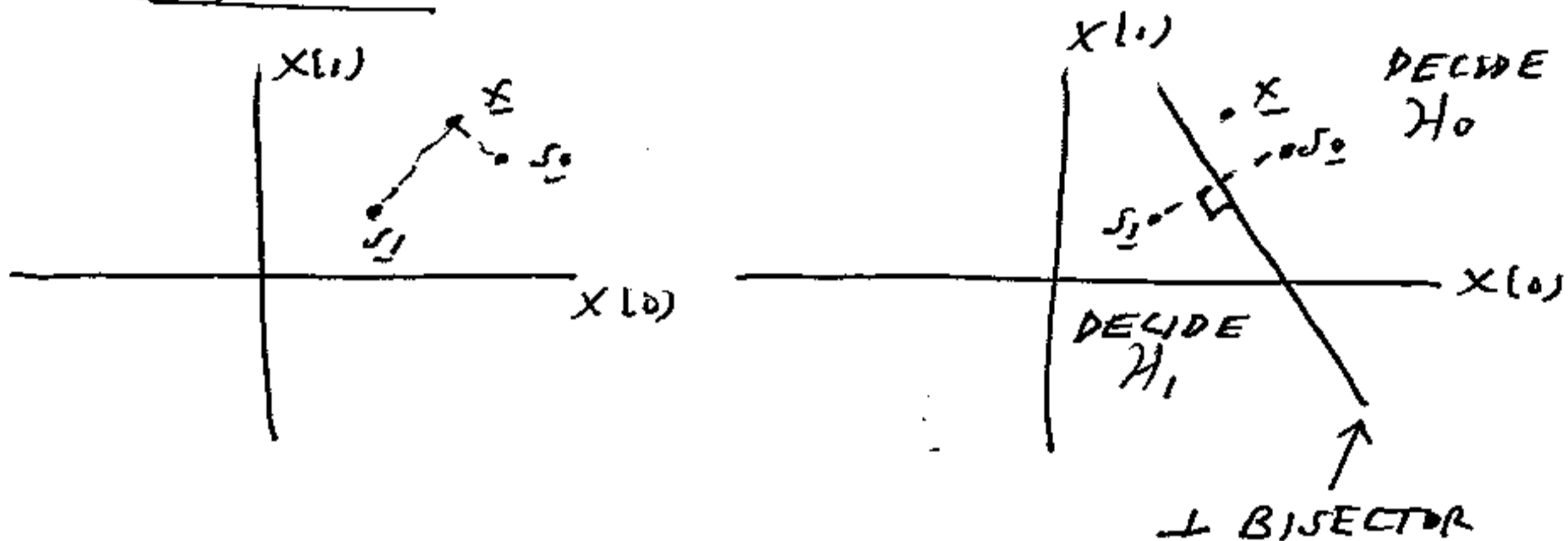
$$D_i^2 = \|x - s_i\|^2$$

$$\text{WHERE } \|z\| = \sqrt{\sum_{i=0}^{N-1} z_i^2}$$

= EUCLIDEAN DISTANCE

CHOOSE s_0 (H_0) IF s_0 IS CLOSER TO x AND OTHERWISE s_1 (H_1).

EXAMPLE : $N=2$



ALTERNATIVE RECEIVER IS

$$D_i^2 = \sum_n (x(n) - s_i(n))^2$$

$$= \sum_n x^2(n) - 2 \sum_n x(n) s_i(n) + E_i$$

TO MINIMIZE $D_i^2 \Rightarrow$ MAXIMIZE

$$T_i(x) = \underbrace{\sum_{n=0}^{N-1} x(n) s_i(n)}_{\text{CORRELATOR}} - \underbrace{\frac{1}{2} E_i}_{\text{BIAS TERM}}$$

IF $E_0 = E_1$, CHOOSE HYPOTHESIS YIELDING LARGER CORRELATION.

PERFORMANCE

WISH TO DETERMINE P_e .

$$\begin{aligned}
 P_e &= P(H_1|H_0)P(H_0) + P(H_0|H_1)P(H_1) \\
 &= \frac{1}{2} [P(H_1|H_0) + P(H_0|H_1)] \\
 &= \frac{1}{2} [P_r \{T_1(\underline{x}) - T_0(\underline{x}) > 0 | H_0\} \\
 &\quad + P_r \{T_0(\underline{x}) - T_1(\underline{x}) > 0 | H_1\}]
 \end{aligned}$$

LET $T = T_1 - T_0$

$$= \sum x(n) (s_1(n) - s_0(n)) - \frac{1}{2} (\epsilon_1 - \epsilon_0)$$

T IS GAUSSIAN

$$\begin{aligned}
 E(T|H_0) &= \sum s_0(n) (s_1(n) - s_0(n)) \\
 &\quad - \frac{1}{2} (\epsilon_1 - \epsilon_0)
 \end{aligned}$$

$$= \sum s_0(n) s_1(n) - \epsilon_0 - \frac{1}{2} \epsilon_1 + \frac{1}{2} \epsilon_0$$

$$= -\frac{1}{2} \sum (s_1(n) - s_0(n))^2$$

$$= -\frac{1}{2} \|\underline{s}_1 - \underline{s}_0\|^2$$

$$\text{SIMILARLY } E(T|H_1) = \frac{1}{2} \|\underline{s}_1 - \underline{s}_0\|^2$$

$$\text{VAR}(T | H_0) = \text{VAR} \left(\sum_{i=1}^n x_i(n) (\underline{\mu}_1 - \underline{\mu}_0) \right) | H_0$$

$$= \sum \text{VAR}(x_i(n)) (\underline{\mu}_1 - \underline{\mu}_0)^2$$

$$= \sigma^2 \|\underline{\mu}_1 - \underline{\mu}_0\|^2 = \text{VAR}(T | H_1)$$

$$\Rightarrow T \sim N \left(-\frac{1}{2} \|\underline{\mu}_1 - \underline{\mu}_0\|^2, \sigma^2 \|\underline{\mu}_1 - \underline{\mu}_0\|^2 \right) | H_0 \\ N \left(\frac{1}{2} \|\underline{\mu}_1 - \underline{\mu}_0\|^2, \sigma^2 \|\underline{\mu}_1 - \underline{\mu}_0\|^2 \right) | H_1$$

$$P_e = \frac{1}{2} P_r \{ T > 0 | H_0 \} + \frac{1}{2} P_r \{ T < 0 | H_1 \}$$

$$= \frac{1}{2} Q \left(\frac{\frac{1}{2} \|\underline{\mu}_1 - \underline{\mu}_0\|^2}{\sqrt{\sigma^2 \|\underline{\mu}_1 - \underline{\mu}_0\|^2}} \right)$$

$$+ \frac{1}{2} \left[1 - Q \left(\frac{-\frac{1}{2} \|\underline{\mu}_1 - \underline{\mu}_0\|^2}{\sqrt{\sigma^2 \|\underline{\mu}_1 - \underline{\mu}_0\|^2}} \right) \right]$$

$$= Q \left(\frac{\frac{1}{2} \|\underline{\mu}_1 - \underline{\mu}_0\|^2}{\sqrt{\sigma^2 \|\underline{\mu}_1 - \underline{\mu}_0\|^2}} \right)$$

SINCE $Q(-x) = 1 - Q(x)$

$$\text{OR } P_e = Q \left(\frac{1}{2} \sqrt{\frac{\|\underline{\mu}_1 - \underline{\mu}_0\|^2}{\sigma^2}} \right)$$

P_e DECREASES AS SEPARATION BETWEEN \underline{s}_0 AND \underline{s}_1 IN \mathbb{R}^N INCREASES.

PRACTICAL CONSTRAINT IS TO CONSTRAIN ENERGY OR AVERAGE ENERGY $\bar{E} = \frac{1}{2}(\epsilon_0 + \epsilon_1)$
($P(H_0) = P(H_1)$)

THEN

$$\begin{aligned} \|\underline{s}_1 - \underline{s}_0\|^2 &= \underline{s}_1^T \underline{s}_1 - 2 \underline{s}_1^T \underline{s}_0 + \underline{s}_0^T \underline{s}_0 \\ &= 2\bar{E} - 2 \underline{s}_1^T \underline{s}_0 \end{aligned}$$

DEFINE $\rho_s = \frac{\underline{s}_1^T \underline{s}_0}{\frac{1}{2}(\underline{s}_1^T \underline{s}_1 + \underline{s}_0^T \underline{s}_0)}$
= SIGNAL CORRELATION

$$\begin{aligned} \rho_s &= \pm 1 & \text{IF } \underline{s}_1 &= \pm \underline{s}_0 \\ &0 & \text{IF } \underline{s}_1^T \underline{s}_0 &= 0 \end{aligned}$$

$$\begin{aligned} \|\underline{s}_1 - \underline{s}_0\|^2 &= 2\bar{E} - 2\bar{E}\rho_s \\ &= 2\bar{E}(1 - \rho_s) \end{aligned}$$

$$\therefore p_e = Q\left(\sqrt{\frac{\bar{\epsilon}(1-\rho_s)}{2\sigma^2}}\right)$$

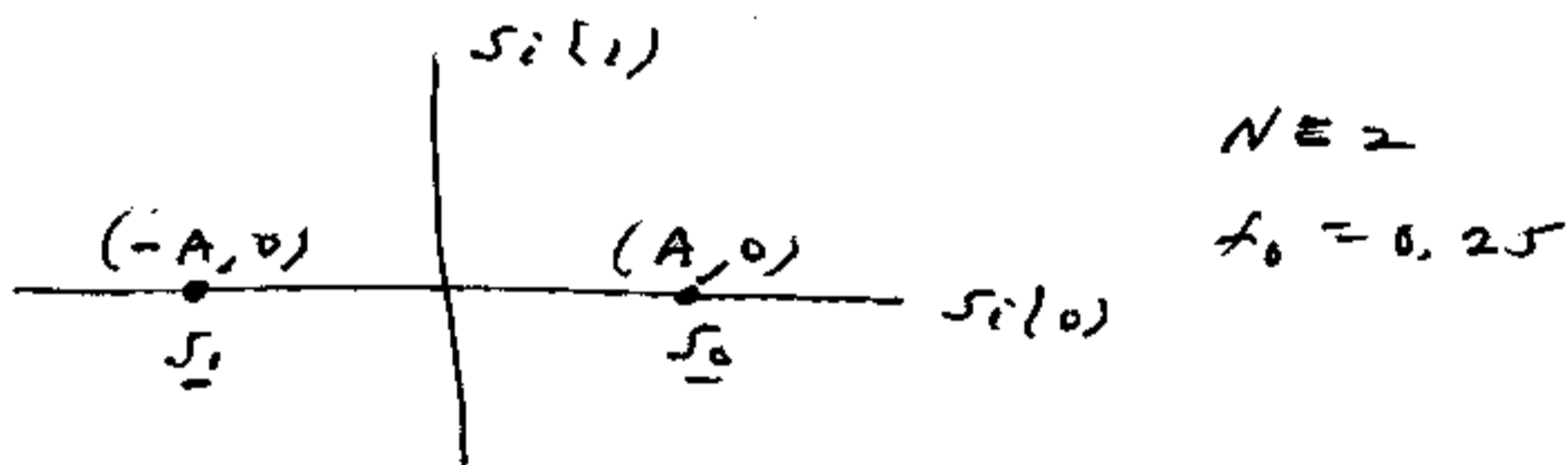
p_e MINIMIZED (FOR GIVEN $\bar{\epsilon}$)
WHEN $\rho_s = -1$ SINCE $|\rho_s| \leq 1$.

EXAMPLES : PHASE SHIFT
KEYING (PSK)

$$s_0(t) = A \cos 2\pi f_0 t$$

$$s_1(t) = A \cos(2\pi f_0 t + \pi) \\ = -A \cos 2\pi f_0 t$$

$$\Rightarrow \underline{s}_1 = -\underline{s}_0$$



$$\underline{s}_1 = -\underline{s}_0 \quad (\text{ANTIPODAL})$$

$\Rightarrow \rho_s = -1 \Rightarrow$ OPTIMAL SIGNALS
MINIMIZES p_e

$$p_e = Q\left(\sqrt{\frac{\bar{\epsilon}}{10\sigma^2}}\right) \quad \bar{\epsilon} = \epsilon_1 = \epsilon_0 \\ \approx NA^2/2$$

FOR $P_e = 10^{-8}$ NEED $\frac{\bar{E}}{\sigma^2} = 15 \text{ dB}$

EXAMPLE : FREQUENCY SHIFT KEYING
(FSK)

$$s_0(n) = A \cos 2\pi f_0 n$$

$$s_1(n) = A \cos 2\pi f_1 n$$

FOR $|f_1 - f_0| \gg 1/N$

$$\sum_n s_0(n) s_1(n) \approx 0 \Rightarrow \rho_{01} \approx 0$$

$$P_e = Q\left(\sqrt{\frac{\bar{E}}{2\sigma^2}}\right)$$

FSK 3 dB POORER THAN PSK
NOT USED MUCH IN PRACTICE

SEE M-ARY CASE

LINEAR MODEL

SEE [KAY 1993] CHAPTER 4

EXAMPLE :

$$x(n) = a \cos 2\pi f_0 n + b \sin 2\pi f_0 n + w(n) \quad n = 0, 1, \dots, N-1$$

\uparrow
 $w(n)$

IN MATRIX FORMAT

$$\underbrace{\begin{bmatrix} x(0) \\ x(1) \\ \vdots \\ x(N-1) \end{bmatrix}}_{\underline{X}} = \underbrace{\begin{bmatrix} 1 & 0 \\ \cos 2\pi f_0 & \sin 2\pi f_0 \\ \vdots \\ \cos 2\pi f_0 (N-1) & \sin 2\pi f_0 (N-1) \end{bmatrix}}_{\underline{H}} \underbrace{\begin{bmatrix} a \\ b \end{bmatrix}}_{\underline{\theta}} + \underbrace{\begin{bmatrix} w(0) \\ w(1) \\ \vdots \\ w(N-1) \end{bmatrix}}_{\underline{W}}$$

H IS $N \times 2$ AND IS KNOWN

θ IS 2×1 AND IS KNOWN

W $\sim N(0, \sigma^2 I)$

EXAMPLE : DC LEVEL IN WHITE GAUSSIAN NOISE

$$x(n) = A + w(n) \quad n = 0, 1, \dots, N-1$$

$$\begin{bmatrix} x(0) \\ x(1) \\ \vdots \\ x(N-1) \end{bmatrix} = \begin{bmatrix} 1 \\ \vdots \\ 1 \end{bmatrix} A + \begin{bmatrix} w(0) \\ w(1) \\ \vdots \\ w(N-1) \end{bmatrix}$$

$\underline{x} \qquad \underline{H} \quad \theta \quad \underline{w}$

$$\underline{w} \sim N(0, \sigma^2 \underline{I})$$

\underline{H}, θ KNOWN

IN GENERAL

$$\underline{x} = \underline{H}\theta + \underline{w} \leftarrow N \times 1$$

$\uparrow \qquad \uparrow \quad \nwarrow$
 $N \times 1 \quad N \times p \quad p \times 1$
 $N > p$

$\underline{w} \sim N(0, \sigma^2 \underline{I})$

OR MORE GENERALLY CAN LET

$$\underline{w} \sim N(0, \underline{C}) \quad \underline{C} \neq \sigma^2 \underline{I}$$

$w(n)$ NOT WGN

WHEN θ IS NOT KNOWN, AN OPTIMAL ESTIMATOR IS

$$\hat{\underline{\theta}} = (\underline{H}^T \underline{C}^{-1} \underline{H})^{-1} \underline{H}^T \underline{C}^{-1} \underline{x}$$

WHICH HAS COVARIANCE MATRIX

$$\underline{C}_{\hat{\theta}} = (\underline{H}^T \underline{C}^{-1} \underline{H})^{-1}$$

(SEE KAY 1993, CHAPTER 4)

WE NOW ASSUME $\underline{\theta}$ IS KNOWN.

EXAMPLE : SINUSOIDAL DETECTION

WISH TO DETECT $s(n) = A \cos(2\pi f_0 n + \phi)$
IN WGN

TO CAST IN LINEAR MODEL FORM:

$$s(n) = \underbrace{A \cos \phi}_{a} \cos 2\pi f_0 n - \underbrace{A \sin \phi}_{b} \sin 2\pi f_0 n$$

$$H_0: x(n) = w(n)$$

$$H_1: x(n) = a \cos 2\pi f_0 n + b \sin 2\pi f_0 n + w(n)$$

FOR $n = 0, 1, \dots, N-1$

$$\text{OR } H_0: \underline{x} = \underline{w}$$

$$H_1: \underline{x} = \underbrace{H\underline{\theta}_1}_{\underline{y}} + \underline{w} \quad \underline{\theta}_1 = \begin{pmatrix} a \\ b \end{pmatrix}$$

THIS IS SAME PROBLEM AS BEFORE

$$\Rightarrow \text{DECIDE } H_1 \text{ IF } T(\underline{x}) = \underline{x}^T \underline{C}^{-1} \underline{y} > \gamma'$$

$$\text{OR } T(\underline{x}) = \underline{x}^T \underline{C}^{-1} H \underline{\theta}_1 > \gamma'$$

FOR THIS EXAMPLE $\underline{C} = \sigma^2 \underline{I}$

$$T(\underline{x}) = \frac{1}{\sigma^2} \underline{x}^T H \underline{\theta}_1$$

$$= \frac{N}{2\sigma^2} \left(\frac{2}{N} H^T \underline{x} \right)^T \underline{\theta}_1$$

$$\text{OR } T'(\underline{x}) = \left(\frac{2}{N} H^T \underline{x} \right)^T \underline{\theta}_1$$

$$\text{BUT } \frac{2}{N} H^T \underline{x} = \begin{bmatrix} \frac{2}{N} \sum_{n=0}^{N-1} x(n) \cos 2\pi f_0 n \\ \frac{2}{N} \sum_{n=0}^{N-1} x(n) \sin 2\pi f_0 n \end{bmatrix}$$

$$\text{AND } (H^T H)^{-1} = \frac{2}{N} \underline{I}$$

$$= \begin{pmatrix} \hat{a} \\ \hat{b} \end{pmatrix} = \text{FOURIER SERIES COEFFICIENTS}$$

$$T'(\underline{x}) = \hat{a} a + \hat{b} b$$

$$= \text{CORRELATION OF TRUE VALUE WITH ESTIMATE}$$

SIGNAL PROCESSING EXAMPLE

PATTERN RECOGNITION - WISH TO CLASSIFY EACH PIXEL OF IMAGE ACCORDING TO GRAY LEVEL

EXAMPLE - BINARY CLASSIFICATION



COULD HAVE M DIFFERENT GRAY LEVELS

ACTUAL GRAY LEVEL DEPENDS ON LIGHTING, ORIENTATION, ETC

MODEL PIXEL VALUES IN A REGION AS RANDOM WITH $x \sim N(\mu_i, \sigma^2 I)$ WHERE ONLY MEAN LEVEL CHANGES

TO MINIMIZE P_e USE MAP RULE
ASSUME $P(H_i) = 1/M \Rightarrow$ USE ML RULE

\Rightarrow MAXIMIZE $p(x | H_i)$ OVER i

WHERE $\underline{x} \sim N(\underline{\mu}_i, \sigma^2 \underline{I})$ $N \times 1$

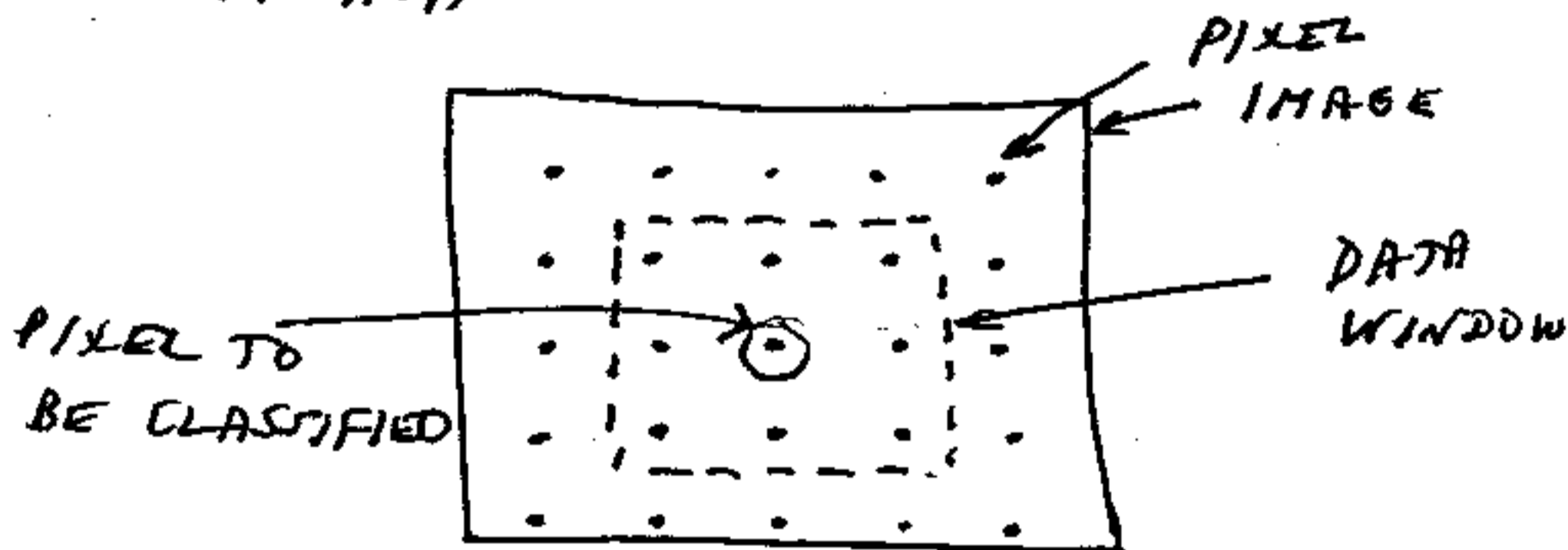
$$p(\underline{x} | H_i) = \frac{1}{(2\pi\sigma^2)^{N/2}} e^{-\frac{1}{2\sigma^2} (\underline{x} - \underline{\mu}_i)^T (\underline{x} - \underline{\mu}_i)}$$

TO MAXIMIZE \Rightarrow MINIMIZE

$$D_i^2 = (\underline{x} - \underline{\mu}_i)^T (\underline{x} - \underline{\mu}_i)$$

$$= \|\underline{x} - \underline{\mu}_i\|^2 \quad \begin{array}{l} \text{MINIMUM} \\ \text{DISTANCE} \\ \text{RECEIVER} \end{array}$$

TO APPLY TO OUR PROBLEM USE FOLLOWING APPROACH



FOR BINARY CLASSIFICATION

ASSUME $\underline{\mu}_0 = 0$ (BLACK) OR

$\underline{\mu}_1 = 1$ (WHITE)

LET \underline{x} = VECTOR OF ALL DATA (PIXEL) VALUES IN WINDOW

$$\begin{aligned} \Rightarrow D_i^2 &= \|\underline{x} - \underline{\mu}_0\|^2 \\ &= \|\underline{x}\|^2 \quad \text{BLACK (H}_0\text{)} \\ &\quad \|\underline{x} - \underline{1}\|^2 \quad \text{WHITE (H}_1\text{)} \end{aligned}$$

AND CHOOSE WHICHEVER IS SMALLER
OR DECIDE H_1 IF

$$\|\underline{x}\|^2 > \|\underline{x} - \underline{1}\|^2$$

$$\begin{aligned} \underline{x}^T \underline{x} &> (\underline{x} - \underline{1})^T (\underline{x} - \underline{1}) \\ &= \underline{x}^T \underline{x} - 2 \underline{1}^T \underline{x} + \underline{1}^T \underline{1} \end{aligned}$$

$$\Rightarrow 2 \underline{1}^T \underline{x} > \underline{1}^T \underline{1} = N$$

$$\frac{1}{N} \underline{1}^T \underline{x} > \frac{1}{2}$$

$\bar{x} > \frac{1}{2}$, AVERAGE PIXELS
IN WINDOW.

NOTE: WE DO THIS FOR ALL PIXELS
ASSUMES THAT ALL PIXELS
WITHIN WINDOW HAVE SAME
MEAN VALUE.

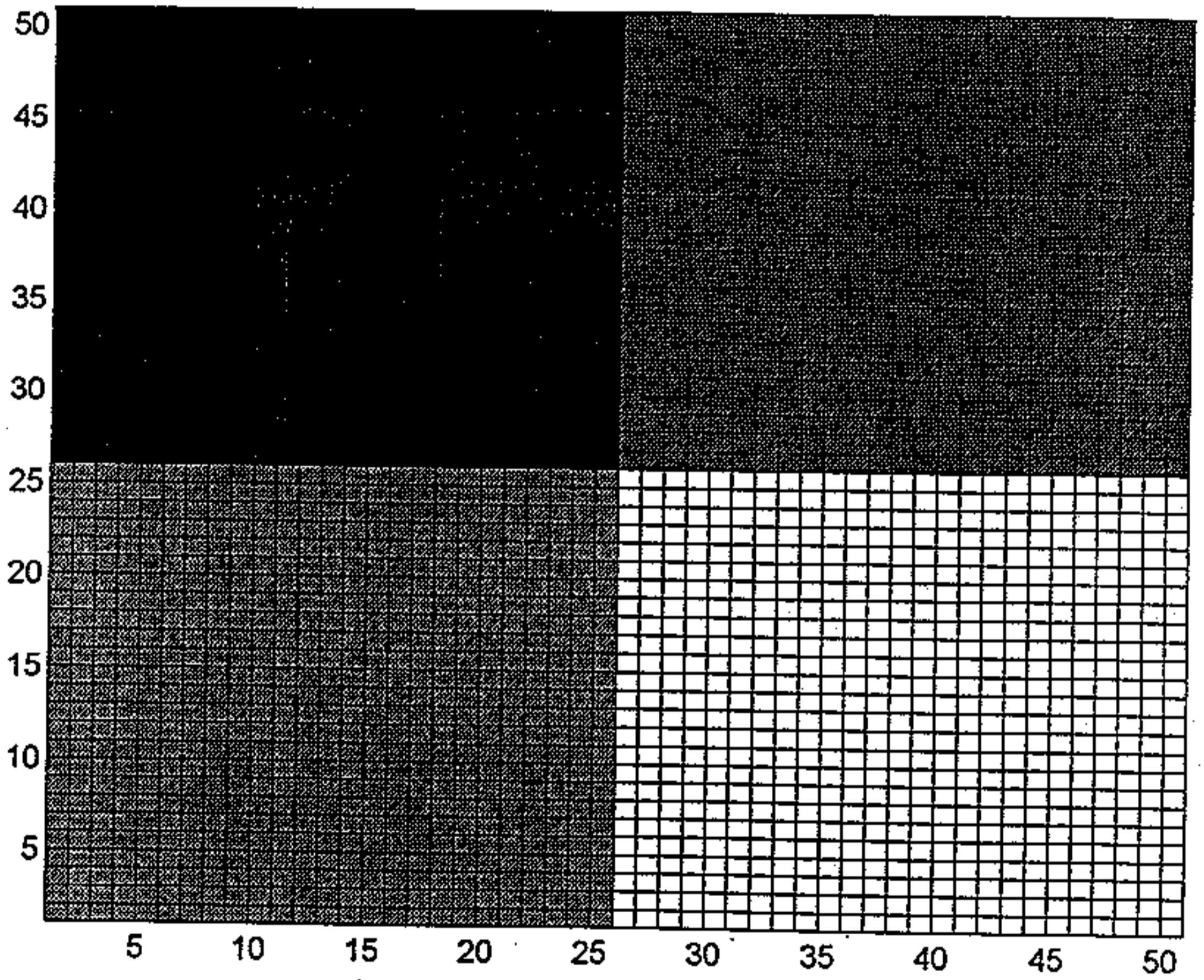


FIGURE 4,17 - SYNTHETIC IMAGE

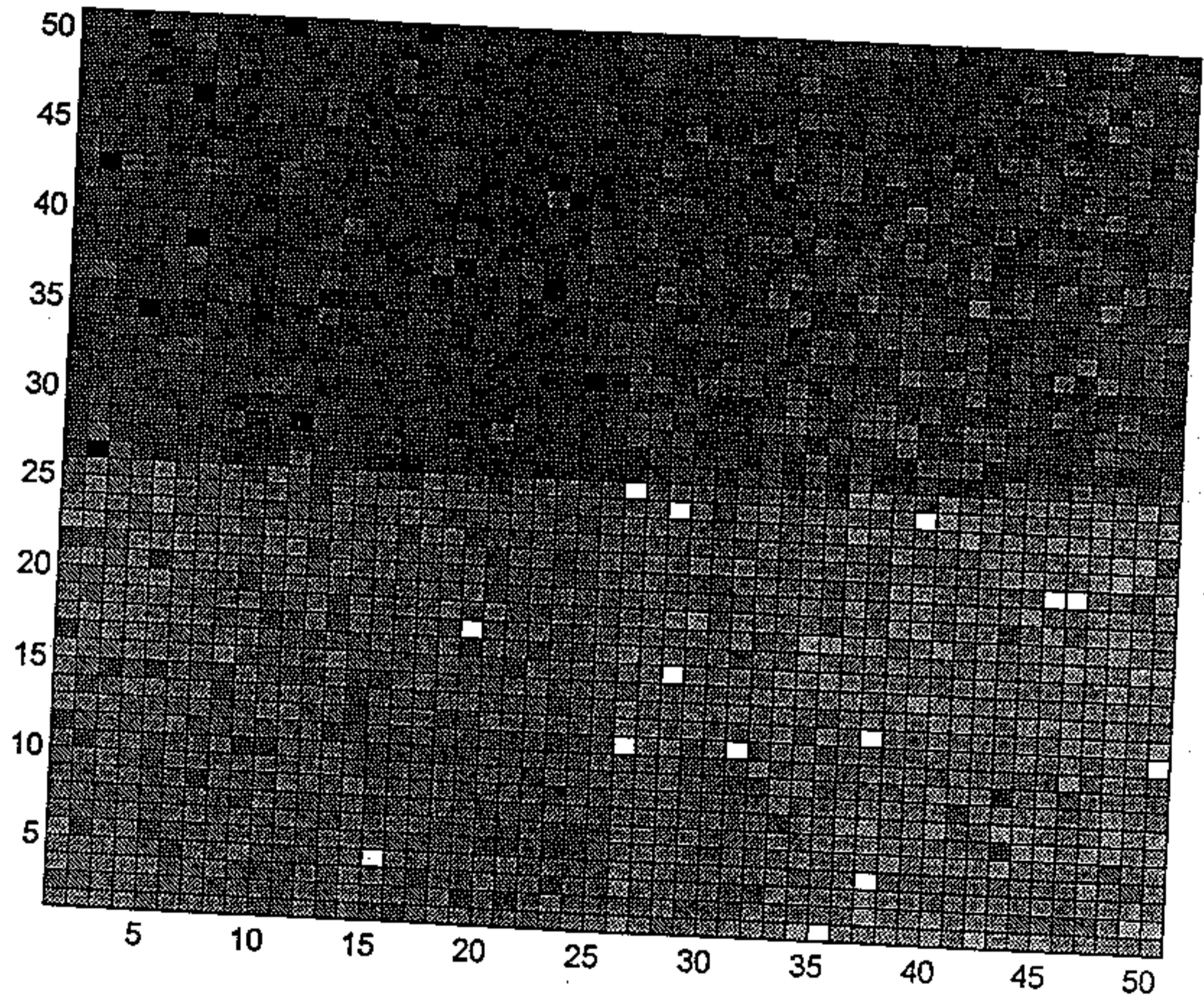


FIGURE 4.18 - NOISE CORRUPTED IMAGE

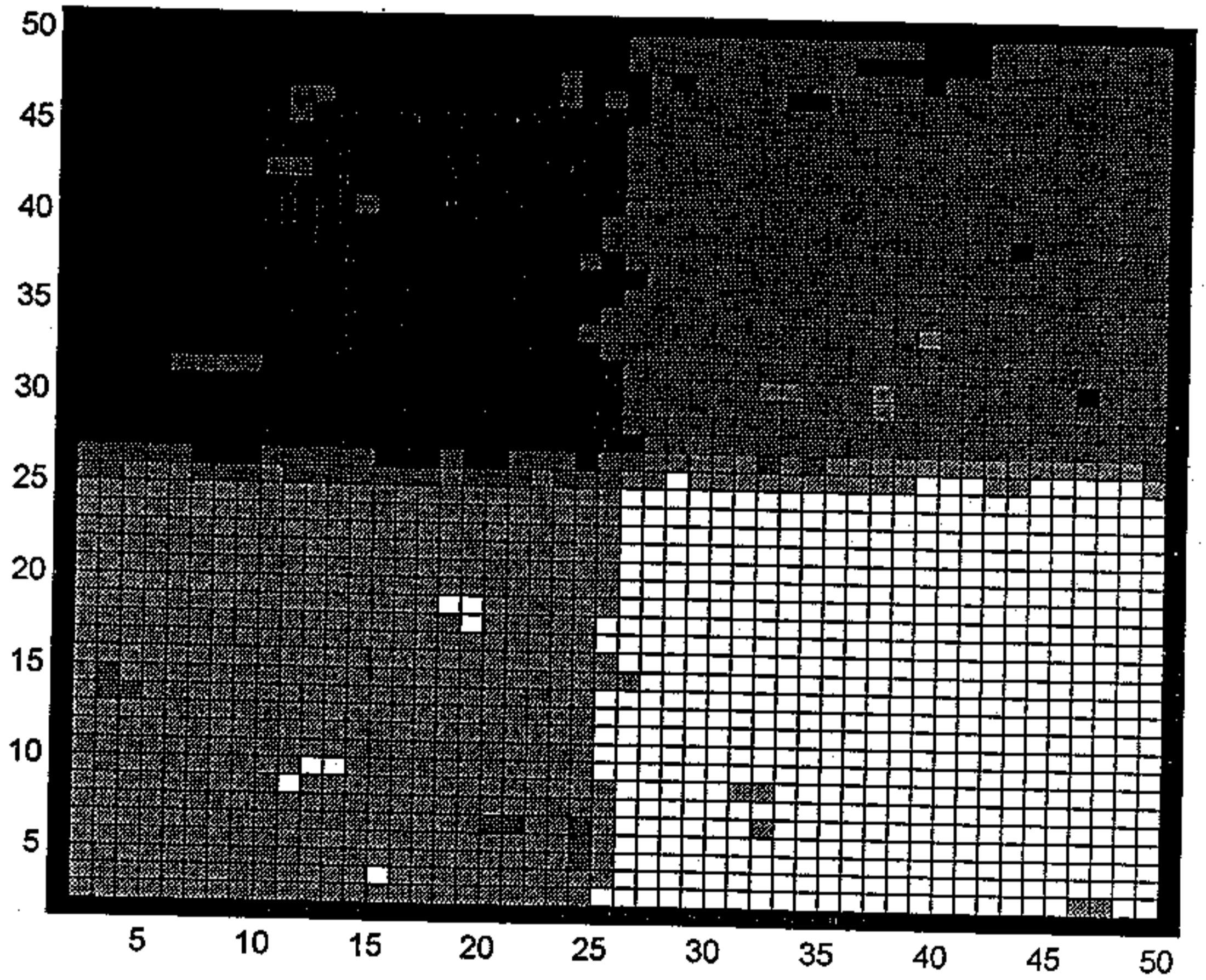


FIGURE 4.19 - CLASSIFICATION USING
3x3 WINDOW

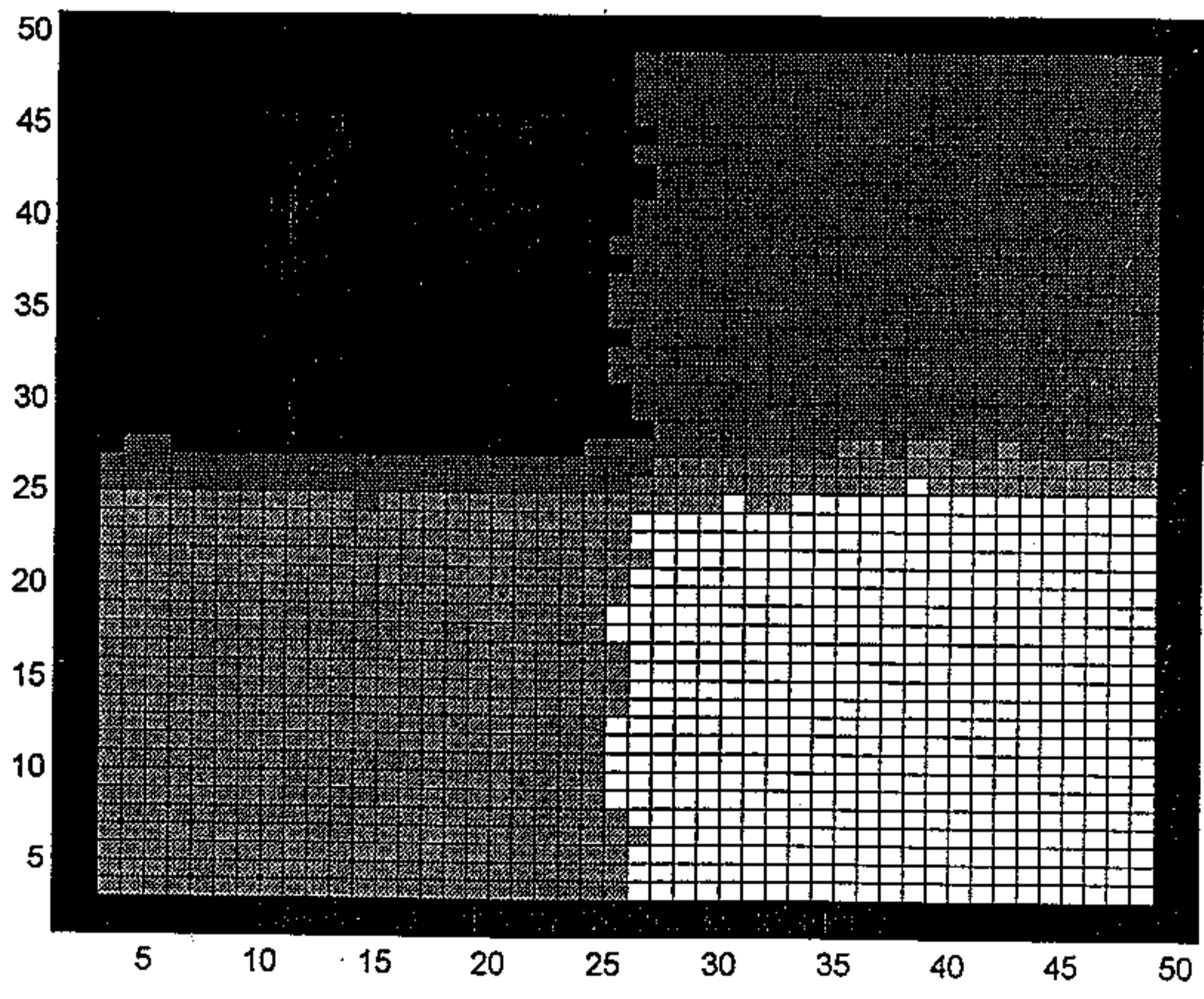


FIGURE 4.20 - CLASSIFICATION USING
5x5 WINDOW

CHAPTER 5 - RANDOM SIGNALS

SOME SIGNALS ARE UNKNOWN IN FORM - CONSIDER A SPEECH SIGNAL

SAME SOUND / SAME SPEAKER \Rightarrow
DIFFERENT WAVEFORMS FOR
EACH UTTERANCE (SEE CHAPTER 1)

USE RANDOM SIGNAL MODEL

ESTIMATOR - CORRELATOR

EXAMPLE: ENERGY DETECTOR

ASSUME SIGNAL IS $s(n)$

WHERE $s(n)$ IS "WGN" WITH
VARIANCE σ_s^2

$$\begin{aligned} H_0: x(n) &= w(n) \\ H_1: x(n) &= s(n) + w(n) \end{aligned}$$

INDEPENDENT OF EACH OTHER

NP DETECTOR COMPUTES

$$L(x) = \frac{p(x; H_1)}{p(x; H_0)}$$

$$\underline{x} = \underline{w} \sim N(\underline{0}, \sigma^2 \underline{I}) \quad H_0$$

$$= \underline{s} + \underline{w} \sim N(\underline{0}, (\sigma^2 + \sigma_s^2) \underline{I}) \quad H_1$$

ONLY DISCRIMINATION IS IN POWER



$$L(\underline{x}) = \frac{1}{[2\pi(\sigma_s^2 + \sigma^2)]^{N/2}} e^{-\frac{1}{2(\sigma_s^2 + \sigma^2)} \sum_n x^2(n)}$$

$$\frac{1}{(2\pi\sigma^2)^{N/2}} e^{-\frac{1}{2\sigma^2} \sum_n x^2(n)}$$

$$l(\underline{x}) = \ln L(\underline{x})$$

$$= \frac{N}{2} \ln \left(\frac{\sigma^2}{\sigma_s^2 + \sigma^2} \right) - \frac{1}{2} \left(\frac{1}{\sigma_s^2 + \sigma^2} - \frac{1}{\sigma^2} \right) \cdot \sum_n x^2(n)$$

$$= \frac{N}{2} \ln \left(\frac{\sigma^2}{\sigma_s^2 + \sigma^2} \right) + \frac{1}{2} \frac{\sigma_s^2}{\sigma^2(\sigma_s^2 + \sigma^2)} \cdot \sum_n x^2(n)$$